

Quiz #3: Ideal op amps and/or Bode plots

Comment on bulk-source aspects of MOS SS. model-

It should be developed using  $N_{BS}$  rather than  $N_{BS}$  (although there should be no difference at the end)

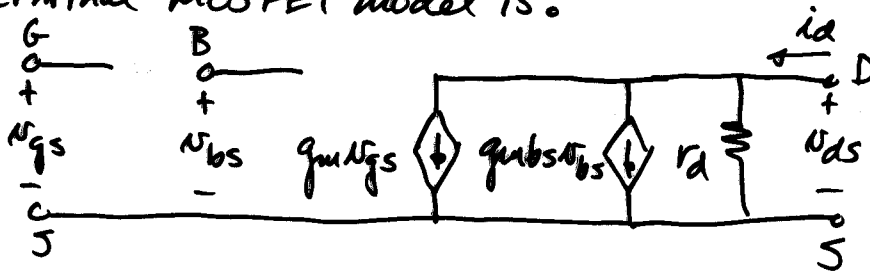
$$V_{TN} = V_{T0} + \gamma \sqrt{2\phi_F - N_{BS}} - \gamma \sqrt{2\phi_F}$$

$$g_{mbs} = \left. \frac{\partial i_D}{\partial N_{BS}} \right|_Q = \left( \frac{\partial i_D}{\partial V_{TN}} \right) \left( \frac{\partial V_{TN}}{\partial N_{BS}} \right) \Big|_Q = (-g_m) \left( \frac{-\gamma}{\sqrt{2\phi_F - V_{BS}}} \right)$$

Normally,  $V_{BS} < 0$  (the bulk-source diode would be forward-biased otherwise)

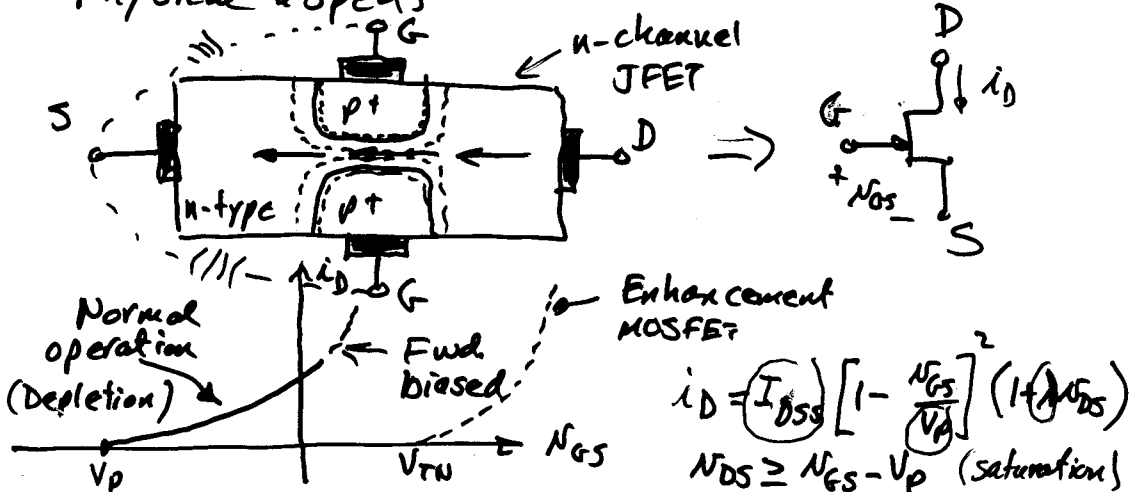
$$\therefore g_{mbs} = \frac{g_m \gamma}{\sqrt{2\phi_F + |V_{BS}|}} = m \gamma \quad (m \approx \frac{1}{10})$$

4-terminal MOSFET model is:



JFET Small Signal Model

Physical aspects



Small signal model -

$$i_D = f(V_{GS}, V_{DS}) \rightarrow i_D = k_1 V_{GS} + k_2 V_{DS} = g_m V_{GS} + g_o V_{DS}$$

$$g_m = \left. \frac{\partial i_D}{\partial V_{GS}} \right|_{V_{DS}=0} \approx \left. \frac{\partial i_D}{\partial V_{GS}} \right|_Q = -\frac{2 I_{DSS}}{V_p} \left(1 - \frac{V_{GS}}{V_p}\right) (1 + \lambda V_{DS})$$

$$g_m = -\frac{2 I_{DSS}}{V_p^2} (V_p - V_{GS}) (1 + \lambda V_{DS}) = \frac{2 I_{DSS}}{V_p^2} (V_{GS} - V_p) (1 + \lambda V_{DS})$$

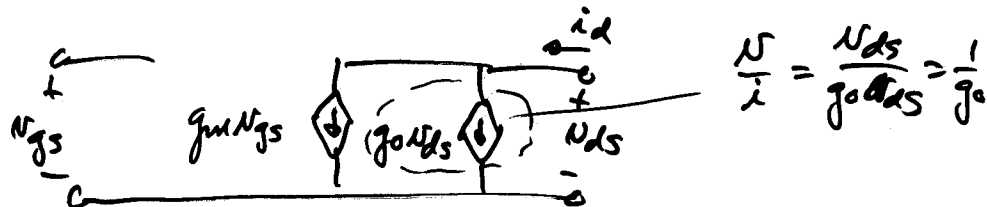
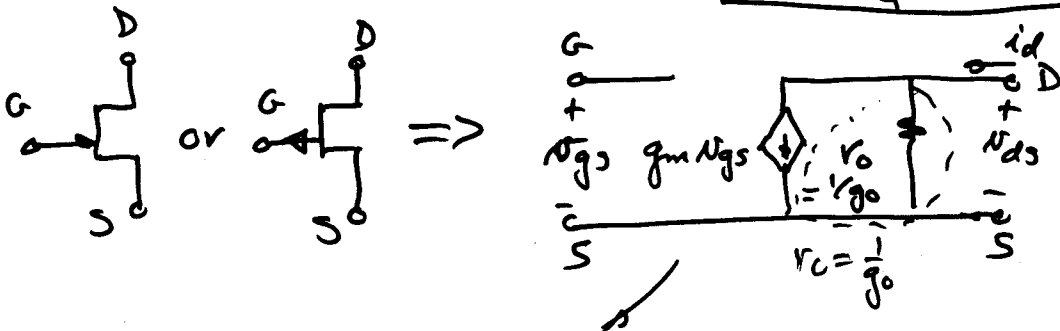
$$I_D = \frac{I_{DSS}}{V_p^2} (V_p - V_{GS})^2 (1 + \lambda V_{DS}) = \frac{I_{DSS}}{V_p^2} (V_{GS} - V_p)^2 (1 + \lambda V_{DS})$$

$$|V_{GS} - V_p| = \sqrt{\frac{I_D V_p^2}{I_{DSS} (1 + \lambda V_{DS})}}$$

$$g_m = \frac{2 I_{DSS}}{V_p^2} (1 + \lambda V_{DS}) \sqrt{\frac{I_D V_p^2}{I_{DSS} (1 + \lambda V_{DS})}} = \frac{2}{|V_p|} \sqrt{I_D I_{DSS}} (1 + \lambda V_{DS})$$

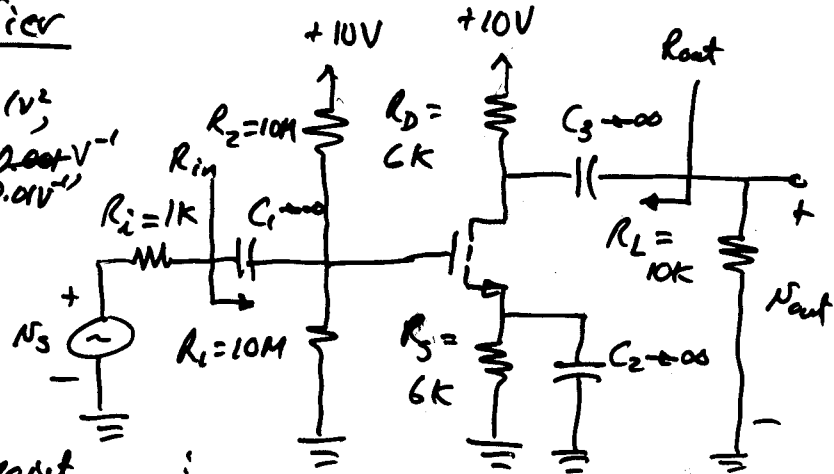
$$g_m \approx \frac{2}{|V_p|} \sqrt{I_D I_{DSS}}$$

$$g_o = \left. \frac{\partial i_D}{\partial V_{DS}} \right|_Q = \frac{\lambda I_D}{1 + \lambda V_{DS}} \approx \lambda I_D$$

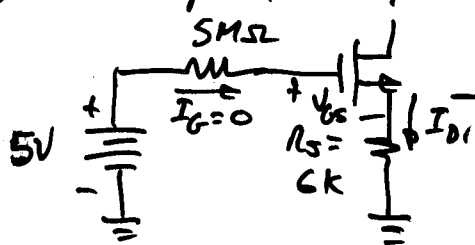


NMOS Amplifier

If  $k_N = 0.5 \text{ mA/V}^2$ ,  
 $V_{TN} = 1 \text{ V}$  and  $\lambda = 0.001 \text{ V}^{-1}$   
 find  $\frac{N_{out}}{N_{in}}$ ,  
 $R_{in}$  and  $R_{out}$ .



a.) Find Q point



$$\begin{cases} 5 = V_{GS} + I_D 6k \\ I_D = \frac{k_N}{2} (V_{GS} - V_{TN})^2 \end{cases}$$

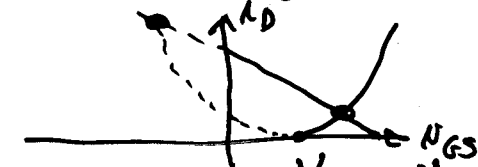
$$V_{GS}^2 - \frac{5}{3} V_{GS} - \frac{2}{3} = 0$$

$$V_{GS} = \frac{5}{6} \pm 1.667 = \underline{2 \text{ V}}$$

$$I_D = \frac{0.5 \text{ mA}}{2} (2 - 1)^2 = \underline{0.5 \text{ mA}}$$

$$V_{DS} = 10 - I_D 6k = 4 \text{ V} \rightarrow \text{check sat. } V_{DS} \geq V_{GS} - V_{TN}$$

$$4 \geq 2 - 1 = 1$$

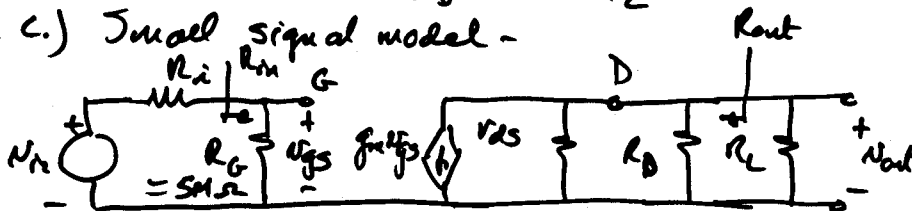


b.) Small signal model parameters

$$g_m = \sqrt{2k_N I_D} (1 + \lambda V_{DS}) = \sqrt{2 \cdot \frac{1}{2} \cdot \frac{1}{2} \cdot (1.04)} = \underline{0.721 \text{ mS}}$$

$$r_{ds} = r_o = \frac{V_{DS} + \frac{1}{\lambda}}{I_D} = \frac{4 + 100}{\frac{1}{2}} = \underline{208 \text{ k}\Omega}$$

c.) Small signal model -



$$R_{in} = R_G = R_1 || R_2 = \underline{5 \text{ M}\Omega}$$

$$R_{out} = r_{ds} || R_D = 6k || 208k = \underline{5.834 \text{ k}\Omega}$$

$$\frac{N_{out}}{N_{in}} = \left( \frac{N_{out}}{N_{gs}} \right) \left( \frac{N_{gs}}{N_{in}} \right) = \left[ -g_m (r_{ds} || R_D || R_L) \right] \left[ \frac{R_G}{R_i || R_G} \right] = (-2.13)(0.999) = \underline{-2.13 \text{ V/V}}$$