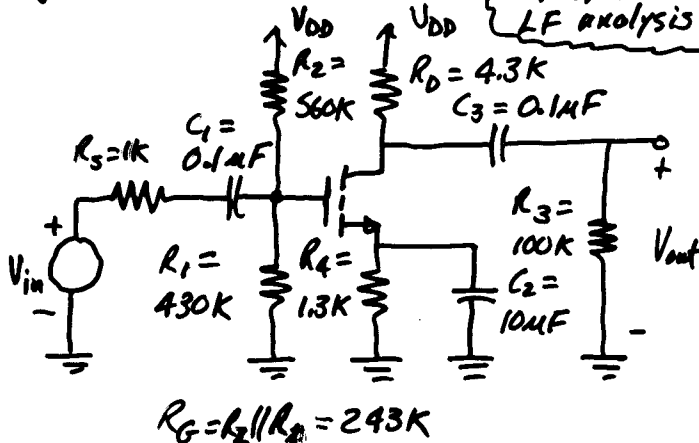


MOSFET Example of High Frequency Analysis

Quiz 8
10/29/04
LF analysis

Find ω_H of the MOSFET amplifier shown if $g_m = 1.23 \text{ mS}$, $C_{gs} = 10 \text{ pF}$, and $C_{gd} = 2 \text{ pF}$
Note: This is the same circuit used for LF analysis (see p. 4 of 10/13/04 lecture)

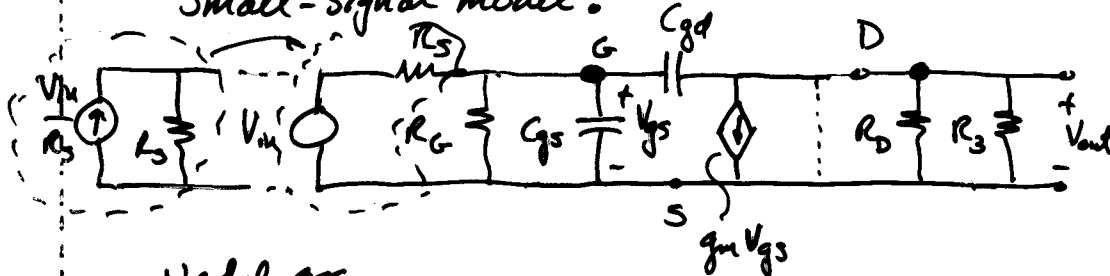


$R_G = R_1 || R_2 = 243k$

$X_C = \frac{1}{\omega C}$

① Direct Analysis

Small-signal model:



Nodal eqs.

$$\left(\frac{1}{R_S} + \frac{1}{R_G} + sC_{gs} + sC_{gd}\right)V_{gs} - sC_{gd}V_{out} = \frac{V_{in}}{R_S}$$

$$\frac{1}{R_S} = G_S \quad sC_{gd}(V_{out} - V_{gs}) + g_m V_{gs} + \left(\frac{1}{R_D} + \frac{1}{R_3}\right)V_{out} = 0$$

Two pages later

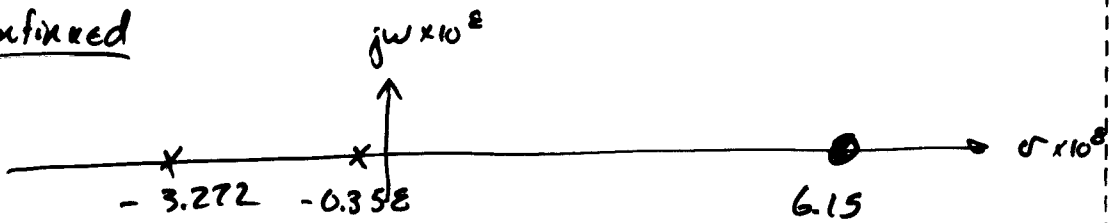
$$\frac{V_{out}}{V_{in}} = \frac{-G_S(g_m - sC_{gd})}{(G_S + G_G)(G_D + G_3) + s[C_{gs}(G_D + G_3) + C_{gd}(G_D + G_3 + G_S + G_G) + g_m C_{gd}]}$$

Substituting the values:

$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{-10^{-3} (1.23 \times 10^{-3} - s 2 \text{ pF})}{20 \times 10^{-24} [s^2 + 3.628 \times 10^6 s + 1.165 \times 10^{16}]}$$

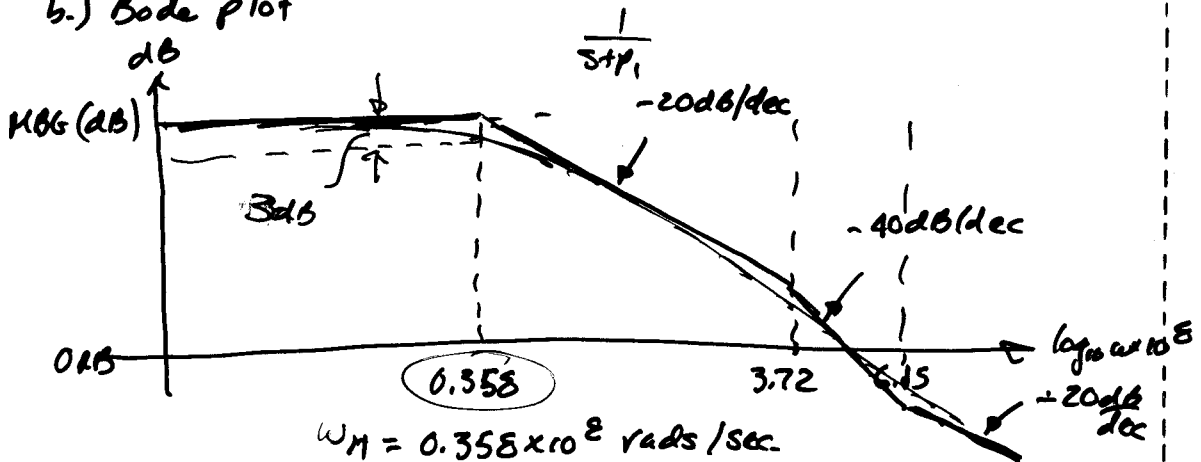
$\omega_1 = -0.358 \times 10^6 \frac{\text{rads}}{\text{sec}}$ & $\omega_2 = -3.272 \times 10^8 \frac{\text{rads}}{\text{sec}}$

Continued

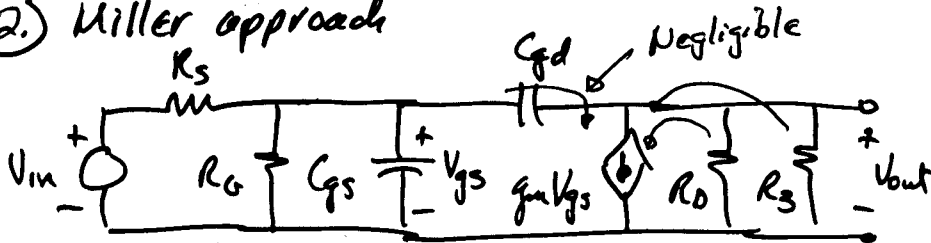


a.) Dominant pole
 $\omega_H \approx |\text{dominant pole}| = 0.358 \times 10^8 \frac{\text{rads}}{\text{sec}} \Rightarrow f_H = 5.67 \text{ MHz}$

b.) Bode plot



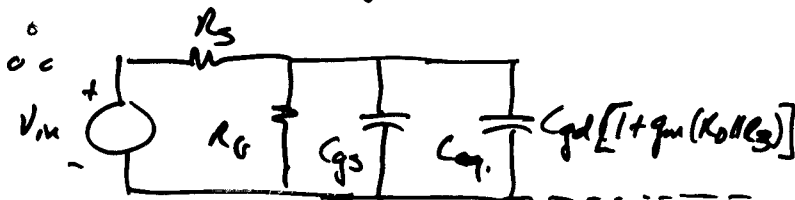
② Miller approach



$$K = \frac{V_{out}}{V_{gs}} = ? \quad (eq. = C_{gd}(1-K))$$

$$V_{out} \approx -g_m(R_D || R_S) V_{gs} \quad \text{if } \frac{1}{\omega_H C_{gd}} \gg R_D || R_S$$

$$C_{eq.} = C_{gd} [1 + g_m(R_D || R_S)]$$



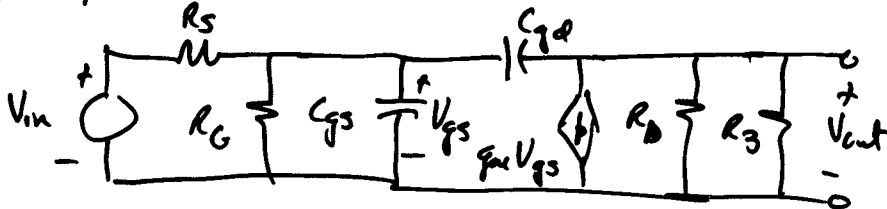
$$\omega_H \approx \frac{1}{[R_S || R_G] [C_{gs} + C_{eq.}]} = \frac{1}{(0.996k) [6pF + 2pF \cdot 6]} = 45.3 \times 10^6 \frac{\text{rads}}{\text{sec}}$$

Cont'd

$$\frac{\omega_H}{2\pi} = f_H = 7.22 \text{ MHz} \quad \frac{1}{\omega_H C_{gd}} = \frac{1}{45.3 \times 10^6 \cdot 2 \text{ pF}} = 11.037 \times 10^{-3}$$

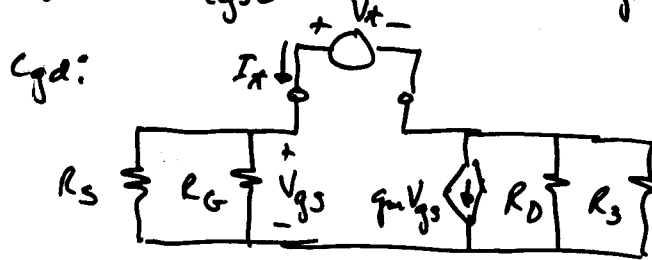
$$R_D || R_3 \approx 4 \text{ K} \quad \frac{1}{\omega_H C_{gd}} > 4 \text{ K}$$

③ Open-Circuit Time Constant



$$\omega_H = \frac{1}{\sum_i R_i C_i}$$

$$C_{gs}: R_{cgs0} = R_S || R_G \rightarrow R_{cgs0} C_{gs} = 0.996 \text{ K} \times 2 \text{ pF}$$



$$V_x = \underbrace{I_x (R_S || R_G)}_{V_{gs}} + (I_x + g_m V_{gs}) (R_D || R_3) = I_x \left\{ R_D || R_3 + R_S || R_G [1 + g_m (R_D || R_3)] \right\}$$

$$R_{cgd0} = \frac{V_x}{I_x} = R_D || R_3 + R_S || R_G [1 + g_m (R_D || R_3)]$$

$$= 4.122 \text{ K} + 0.996 \text{ K} [1 + 5] = 10.16 \text{ K}$$

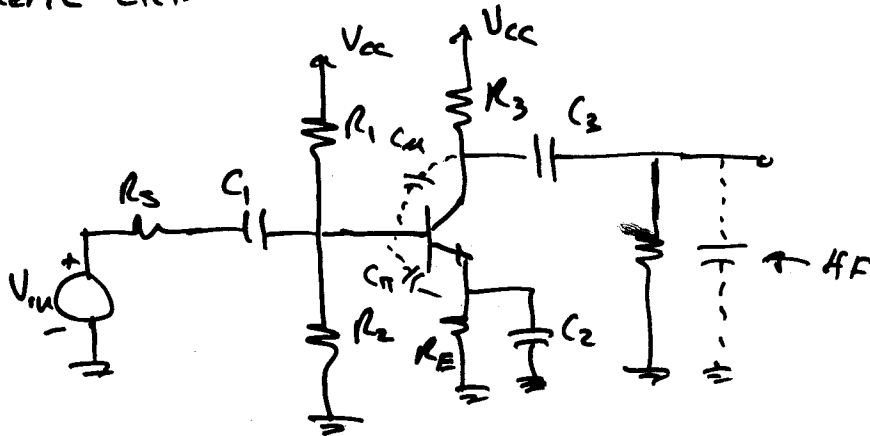
$$\omega_H \approx \frac{1}{R_{cgs0} C_{gs} + R_{cgd0} C_{gd}} = \frac{1}{(0.996 \text{ K})(2 \text{ pF}) + (10.16 \text{ K})(2 \text{ pF})}$$

$$\approx 33.33 \times 10^6 \frac{\text{rads}}{\text{sec}} \rightarrow \underline{f_H = 5.3 \text{ MHz}}$$

There will be 2 examples of HF analysis on web site

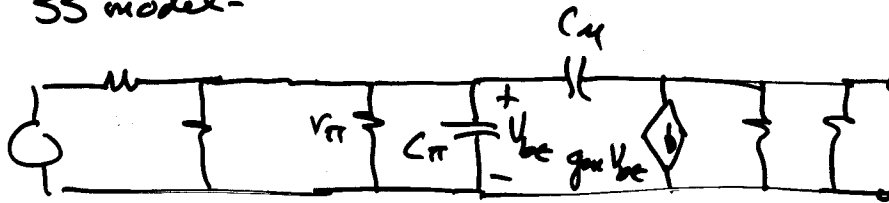
BJT HF Analysis

Generic ckt.



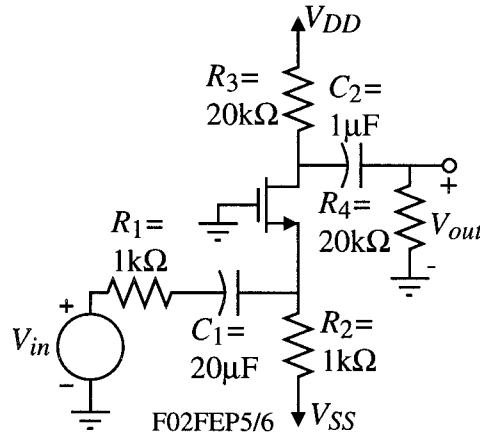
Given: $g_m, v_{\pi}, C_{\pi} \text{ \& } C_{\mu}$

HF SS model -



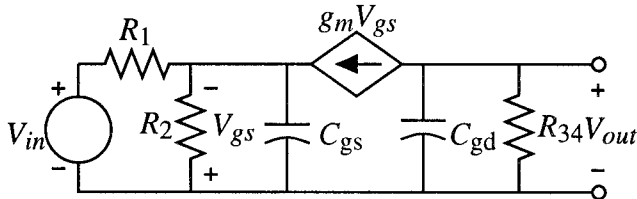
Problem 6 - (20 points - This problem is optional)

The FET in the amplifier shown has $g_m = 1\text{mA/V}$, $r_d = \infty$, $C_{gd} = 0.5\text{pF}$, and $C_{gs} = 10\text{pF}$.
(a.) Find the midband gain, V_{out}/V_{in} . (b.) Find the upper -3dB frequency, f_H , in Hz. (Note: You cannot use the Miller's theorem on this problem because there is no bridging capacitor.)



Solution

The small signal model for the high frequency range is shown where $R_{34} = R_3 \parallel R_4 = 10\text{k}\Omega$.



Because the capacitors are independent, probably the best way to work this problem is by superposition (open-circuit time constants). Therefore,

C_{gs} :

$$R_{C_{gs}} = R_1 \parallel R_2 \parallel (1/g_m) = 1\text{K} \parallel 1\text{K} \parallel 1\text{K} = 333\Omega \rightarrow \omega_{C_{gs}} = \frac{1}{C_{gs} \cdot 333\Omega} = 300 \text{ Mrads/sec.}$$

C_{gd} :

$$R_{C_{gd}} = R_{34} = 10\text{k}\Omega \rightarrow \omega_{C_{gd}} = \frac{1}{C_{gd} \cdot 10\text{k}\Omega} = 200 \text{ Mrads/sec.}$$

$$\therefore \omega_H \approx \frac{1}{\sqrt{\left(\frac{1}{300\text{Mrads/sec}}\right)^2 + \left(\frac{1}{200\text{Mrads/sec}}\right)^2}} = 166 \text{ Mrads/sec.}$$

$$f_L = 26.48\text{MHz}$$

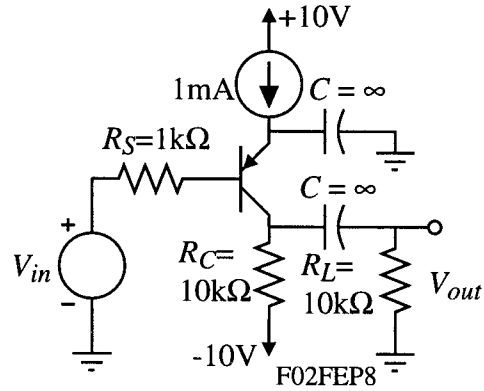
The midband gain is given as

$$\text{MBG} = \left(\frac{R_2 \parallel \frac{1}{g_m}}{R_1 + R_2 \parallel \frac{1}{g_m}} \right) \left(\frac{-g_m R_3 R_4}{R_3 + R_4} \right) = \left(\frac{-0.5}{1.5} \right) (-10) = 3.33 \text{ V/V}$$

Problem 8 – (20 points, this problem is optional)

A common-emitter BJT amplifier is shown. Assume that the BJT has a $\beta = h_{fe} = 100$, $C_{\mu} = 2\text{pF}$, $V_t = 25\text{mV}$, $f_T = 500\text{MHz}$, $r_b = 0\Omega$, and $r_o = \infty$.

- a.) Find the numerical values of r_{π} , g_m , and C_{π} .
- b.) If $r_{\pi} = 1\text{k}\Omega$, $g_m = 0.01\text{A/V}$ and $C_{\pi} = 10\text{pF}$ for the above amplifier, find the value of the upper -3dB frequency, f_H , in Hz.



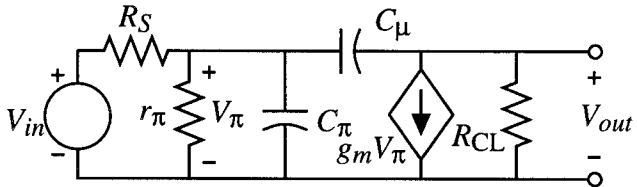
Solution

a.)
$$g_m = \frac{I_C}{V_T} = \frac{1\text{mA}}{25\text{mV}} = 0.04 \text{ A/V}$$

$$r_{\pi} = \frac{\beta_o}{g_m} = \frac{100}{0.04} = 2500\Omega$$

$$C_{\pi} = \frac{g_m}{\omega_T} - C_{\mu} = \frac{0.04}{2\pi \cdot 500 \times 10^6} - 2\text{pF} = 12.732\text{pF} - 2\text{pF} = 10.732 \text{ pF}$$

- b.) The high-frequency, small-signal model for this problem is shown where $R_{CL} = R_C \parallel R_L = 5\text{k}\Omega$. The midband gain of this amplifier is given by



$$\frac{V_{out}}{V_{in}} = \left(\frac{V_{out}}{V_{\pi}}\right) \left(\frac{V_{\pi}}{V_{in}}\right) = -g_m R_C \parallel R_L \left(\frac{r_{\pi}}{r_{\pi} + R_S}\right) = (-0.01 \cdot 5\text{k}\Omega)(0.5) = -25\text{V/V}$$

$$\therefore \text{MBG} = -25 \text{ V/V}$$

Using Miller's theorem on this problem:

If $\frac{1}{\omega_H C_{\mu}} \gg R_C \parallel R_L$, then $C_{eq} \approx C_{\pi} + C_{\mu}(1 + g_m R_C \parallel R_L) = 10\text{pf} + 2\text{pF}(1+50) = 112\text{pF}$

We know that, $\omega_H = \frac{1}{C_{eq}(r_{\pi} \parallel R_S)} = \frac{1}{(112\text{pF} \cdot 500\Omega)} = 17.86 \text{ Mrads/sec.}$

$$\therefore f_H = \frac{\omega_H}{2\pi} = 2.842 \text{ MHz}$$

Note that:

$$\frac{1}{\omega_H C_{\mu}} = \frac{10^6}{17.86 \cdot 2} = 28.06\text{k}\Omega > 5\text{k}\Omega$$
 so that the Miller approximation (neglecting C_{μ})

is valid.