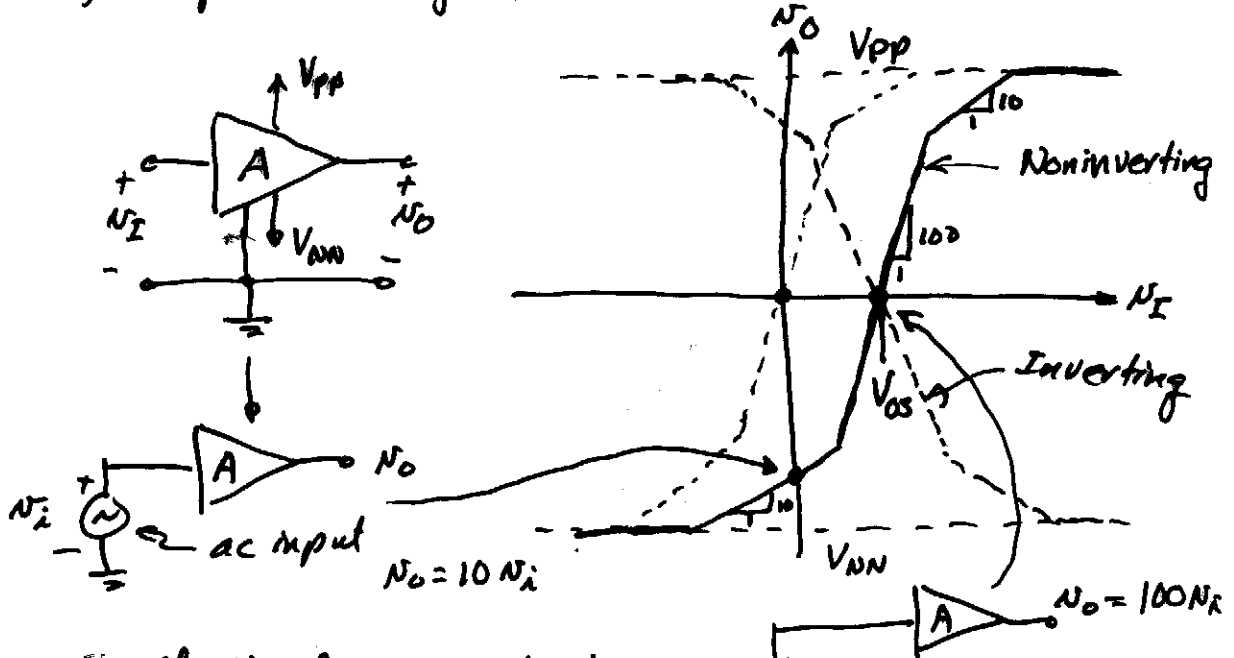
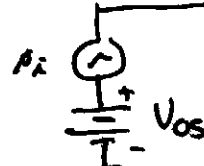


Amplifiers - Chapter 11

1.) Amplifier Voltage Transfer Characteristic

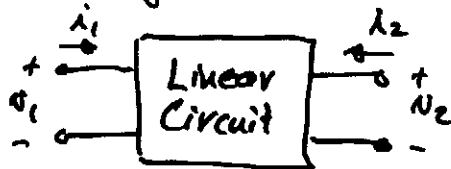


Small-signal Gain  $\equiv A_N = \left. \frac{dN_O}{dN_I} \right|_Q$



2.) Two-port network models for linear amplifiers.

a.) g-parameters

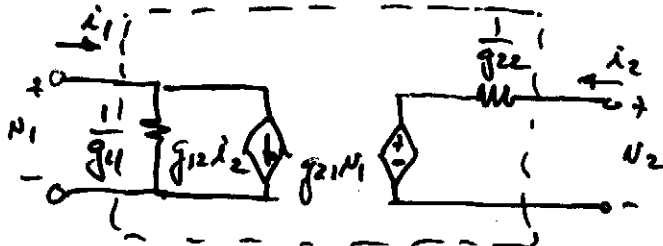


$$i_1 = g_{11} v_1 + g_{12} i_2$$

$$v_2 = g_{21} v_1 + g_{22} i_2$$

$$g_{11} = \left. \frac{i_1}{v_1} \right|_{i_2=0}, \quad g_{12} = \left. \frac{i_1}{i_2} \right|_{v_1=0}, \quad g_{21} = \left. \frac{v_2}{v_1} \right|_{i_2=0}, \quad g_{22} = \left. \frac{v_2}{i_2} \right|_{v_1=0}$$

Schematic model -



b) h-parameter

$$N_1 = h_{11} i_1 + h_{12} N_2$$

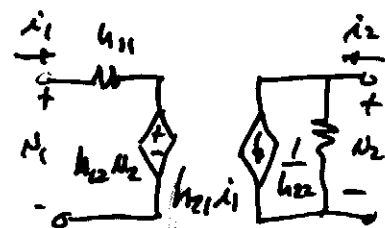
$$i_2 = h_{21} i_1 + h_{22} N_2$$

$$h_{11} = \left. \frac{N_1}{i_1} \right|_{N_2=0}$$

$$h_{12} = \left. \frac{N_1}{N_2} \right|_{i_1=0}$$

$$h_{21} = \left. \frac{i_2}{i_1} \right|_{N_2=0}$$

$$h_{22} = \left. \frac{i_2}{N_2} \right|_{i_1=0}$$



c) g-parameters

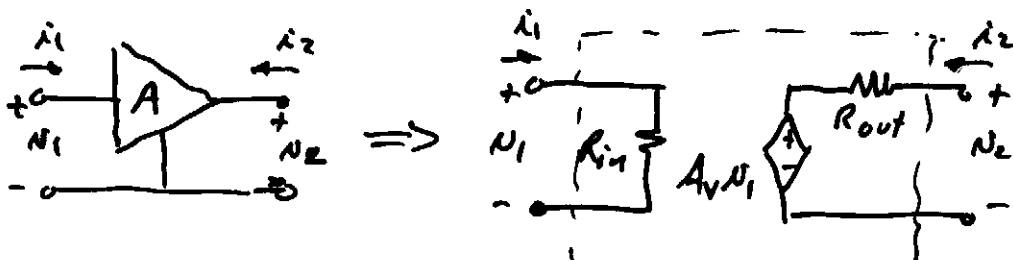
d) y-parameters

3) Back to voltage amplifiers

Essentially any amplifier is a circuit where

$$P_{12} \ll P_{21} \quad \text{where } p = h, g, y, \text{ or } z$$

A voltage amplifier can be modelled as follows,



Find the g-parameters for this voltage amplifier.

$$i_1 = g_{11} N_1 + g_{12} i_2$$

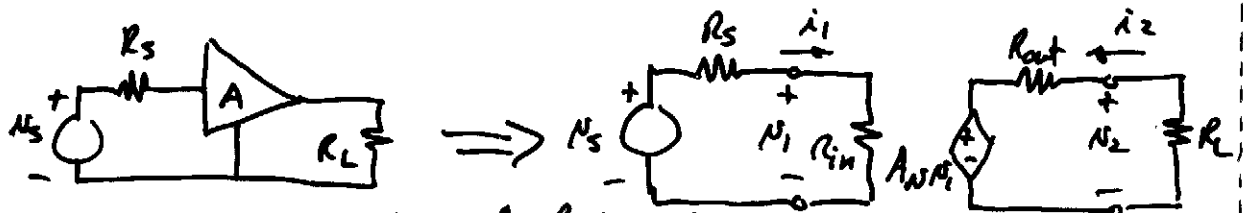
$$N_2 = g_{21} N_1 + g_{22} i_2$$

$$g_{11} = \left. \frac{i_1}{N_1} \right|_{i_2=0} = \frac{1}{R_{in}} \quad g_{12} = \left. \frac{i_1}{i_2} \right|_{N_1=0} = 0$$

$$g_{21} = \left. \frac{N_2}{N_1} \right|_{i_2=0} = A_v$$

$$g_{22} = \left. \frac{N_2}{i_2} \right|_{N_1=0} = R_{out}$$

Loaded amplifier-



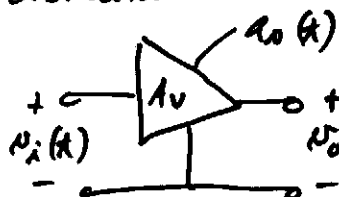
$$\frac{N_2}{N_s} = ? = \left( \frac{N_2}{N_1} \right) \left( \frac{N_1}{N_s} \right) = \left( \frac{A_v R_L}{R_{out} + R_L} \right) \left( \frac{R_{in}}{R_s + R_{in}} \right) = \frac{A_v R_L R_{in}}{(R_{out} + R_L)(R_s + R_{in})} = A_v$$

If  $R_L \rightarrow \infty$  and  $R_S \rightarrow 0$ , then  $A_V \rightarrow A_{Vr}$

## FREQUENCY RESPONSE OF AMPLIFIERS

Assuming our amplifier is time-dependent, then how do you characterize the amplifier?

Time domain -



Convolution

$$v_o(t) = v_i(t) * h_v(t) = \int_0^t v_i(t-\tau) h_v(\tau) d\tau$$

Frequency domain -



$$V_o(j\omega) = A_V(j\omega) V_i(j\omega) \rightarrow A_V(j\omega) = \frac{V_o(j\omega)}{V_i(j\omega)}$$

Complex frequency domain -

$$s = \sigma + j\omega \quad A_V(s) = \frac{V_o(s)}{V_i(s)} \xrightarrow{s=j\omega} A_V(j\omega) = \frac{V_o(j\omega)}{V_i(j\omega)}$$

General Frequency Domain Transfer Function -

$$A_V(s) = \frac{N(s)}{D(s)} = \frac{a_m s^m + a_{m-1} s^{m-1} + \dots + a_1 s + a_0}{b_n s^n + b_{n-1} s^{n-1} + \dots + b_1 s + b_0}$$

$$= K \frac{(s+z_1)(s+z_2) \dots (s+z_m)}{(s+p_1)(s+p_2) \dots (s+p_n)}$$

$z_i$  are called zeros  $[A_V(z_i) = 0]$

$p_k$  are called poles  $\rightarrow A_V(p_k) = \infty$

Example

$$A_V(s) = \frac{100(s^2 + s + 0)}{s^2 + 110s + 1000} = \frac{100 s (s+1)}{(s+10)(s+100)}$$

Zeros are at  $z_1 = 0$  and  $z_2 = -1$  Poles are at  $p_1 = -10$  &  $p_2 = -100$