

Homework Assignment No. 13 - Solutions

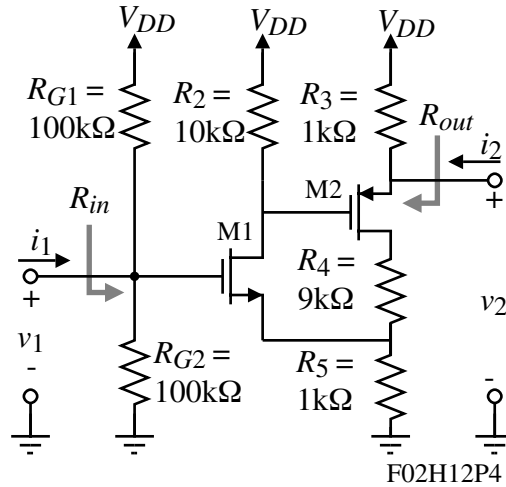
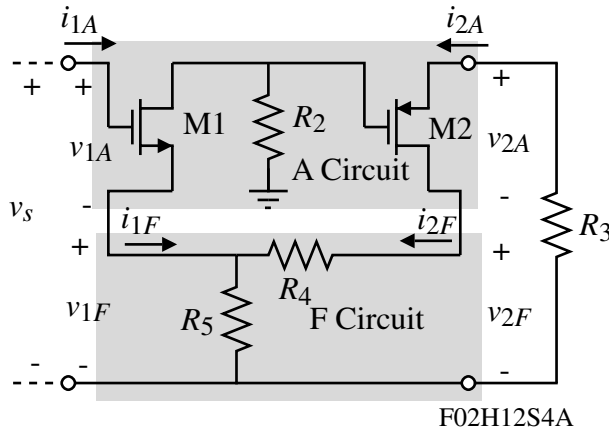
Problem 1

Use the method of feedback analysis to find v_2/v_1 , $R_{in} = v_1/i_1$, and $R_{out} = v_2/i_2$. Assume that all transistors are matched and that $g_{m1} = g_{m2} = 1\text{mS}$. Neglect r_{ds} .

Solution

The topology is series-series.

The circuit is redrawn below in order to help identify the various terms for the analysis:

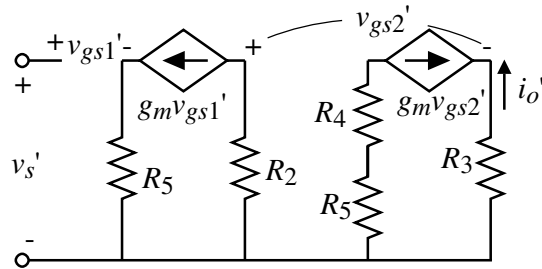
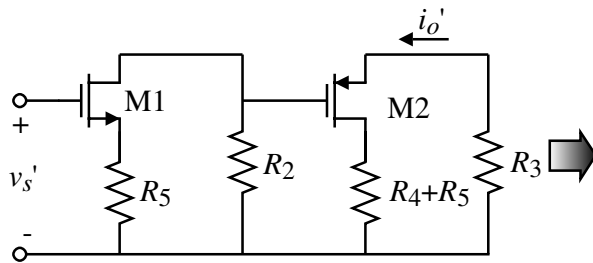


$$z_{11F} = \left. \frac{v_{1F}}{i_{1F}} \right|_{i_{2F}=0} = R_5 = 1\text{k}\Omega$$

$$z_{22F} = \left. \frac{v_{2F}}{i_{2F}} \right|_{i_{1F}=0} = R_4 + R_5 = 10\text{k}\Omega$$

$$z_{12F} = \beta = \left. \frac{v_{1F}}{i_{2F}} \right|_{i_{1F}=0} = R_5 = 1\text{k}\Omega$$

Calculation of the A circuit:



F02H12S4B

$$A = \frac{i_o'}{v_s'} = \left(\frac{i_o'}{v_{gs2'}} \right) \left(\frac{v_{gs2'}}{v_{gs1'}} \right) \left(\frac{v_{gs1'}}{v_s'} \right) = (-g_m) \left(\frac{-g_m R_2}{1 + g_m R_3} \right) \left(\frac{1}{1 + g_m R_5} \right) = (10^{-3})(-5)(-0.5) = 2.5 \text{ mS}$$

$$\therefore \frac{i_o}{v_s} = \frac{A}{1 + A\beta} = \frac{2.5\text{mS}}{1 + 2.5} = 0.714\text{mS}$$

Since, $z_{11A} = \infty$, then $R_{in} = 50\text{k}\Omega \parallel \infty = \underline{50\text{k}\Omega}$

$$\frac{v_2}{v_1} = \frac{-i_o R_3}{v_s} = \frac{-i_o}{v_s} R_3 = -0.714\text{mS}(1\text{k}\Omega) = \underline{-0.714 \text{ V/V}}$$

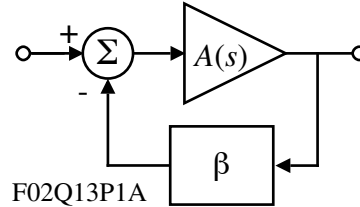
$$R_o = (z_{22T} + R_3)(1 + A\beta) = \left(\frac{1}{g_m} + R_3 \right) (1 + A\beta) = 2\text{k}\Omega(3.5) = 7\text{k}\Omega$$

However, the output resistance specified by the problem, R_{out} , is found as

$$R_{out} = (R_o - R_3) \parallel R_3 = 6\text{k}\Omega \parallel 1\text{k}\Omega = \underline{857\Omega}$$

Problem 2 - The amplifier in the feedback circuit shown has a transfer function of

$$A(s) = \frac{100}{\frac{s}{10^5} + 1}$$



What value of β will increase the upper -3dB frequency by a factor of 10 for the closed loop gain? What is the closed loop, low frequency gain?

Solution

$$A_F = \frac{A}{1+A\beta} = \frac{1}{\frac{1}{A} + \beta} = \frac{1}{\frac{s/10^5 + 1}{100} + \beta} = \frac{100}{\frac{s}{10^5} + 1 + 100\beta} = \left(\frac{100}{1+100\beta}\right) \frac{1}{\frac{s}{10^5(1+100\beta)} + 1}$$

$$\therefore 10^5(1+100\beta) = 10^6 \quad \rightarrow \quad 1+100\beta = 10 \quad \rightarrow \quad \underline{\underline{\beta = 9/100 = 0.09}}$$

The closed-loop, low frequency gain is,

$$A_F(0) = \frac{100}{1+100\beta} = \frac{100}{1+9} = 10 \quad \rightarrow \quad \underline{\underline{A_F(0) = 10}}$$

3.) Problem 18.40 of the text.

$$T = \frac{v_o}{v_x} = g_{m2}(r_{o2} \parallel r_{o4}) \frac{(\beta_o + 1)R}{(r_{o2} \parallel r_{o4}) + r_{\pi3} + (\beta_o + 1)R} \quad | \quad g_{m1} = 40(10^{-4}) = 4.00\text{mS}$$

$$r_{o2} = \frac{50 + 1.4}{10^{-4}} = 514\text{k}\Omega \quad | \quad r_{o4} = \frac{50 + 11.3}{10^{-4}} = 613\text{k}\Omega \quad | \quad r_{\pi3} = \frac{100(0.025)}{(12\text{V}/10\text{k}\Omega)} = 2.08\text{k}\Omega$$

$$T = (4 \times 10^{-3})(280\text{k}\Omega) \frac{(101)10\text{k}\Omega}{280\text{k}\Omega + 2.08\text{k}\Omega + 101(10\text{k}\Omega)} = 876 \quad (58.9\text{ dB})$$

4.) Problem 18.59 of the text.

$$(a) \quad A(s) = \frac{2 \times 10^{14} \pi^2}{(2\pi \times 10^3)(2\pi \times 10^5)} = \frac{5 \times 10^5}{\left(1 + \frac{s}{2\pi \times 10^3}\right) \left(1 + \frac{s}{2\pi \times 10^5}\right)} = \frac{5 \times 10^5}{\left(1 + \frac{s}{2\pi \times 10^3}\right) \left(1 + \frac{s}{2\pi \times 10^5}\right)}$$

$A(s)$ represents a low - pass amplifier with two widely - spaced poles

Open - loop : $A_o = 5 \times 10^5 = 114\text{dB}$ | $f_L = 0$ | $f_H \cong f_1 = 1000\text{ Hz}$

(b) A common mistake would be the following :

$$\text{Closed - loop : } f_H = 1000\text{Hz} [1 + 5 \times 10^5 (0.01)] = 5\text{MHz}$$

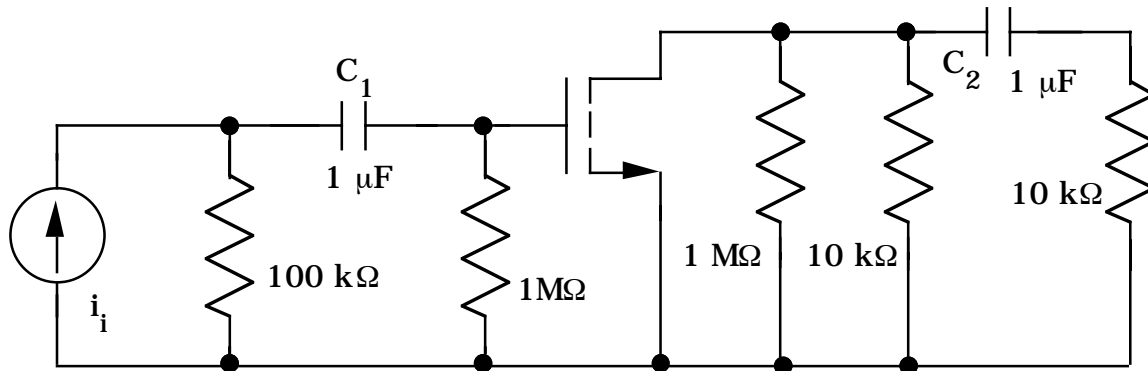
Oops! - This exceeds $f_2 = 100\text{ kHz}$! This is a two - pole low - pass amplifier.

$$A_v(s) = \frac{2 \times 10^{14} \pi^2}{\left(1 + \frac{s}{2\pi \times 10^3}\right) \left(1 + \frac{s}{2\pi \times 10^5}\right) (0.01)} = \frac{2 \times 10^{14} \pi^2}{s^2 + 1.01(2\pi \times 10^5)s + 2 \times 10^{12} \pi^2}$$

Using dominant - root factorization : $f_1 = 101\text{ kHz}$, $f_2 = 4.95\text{ MHz}$

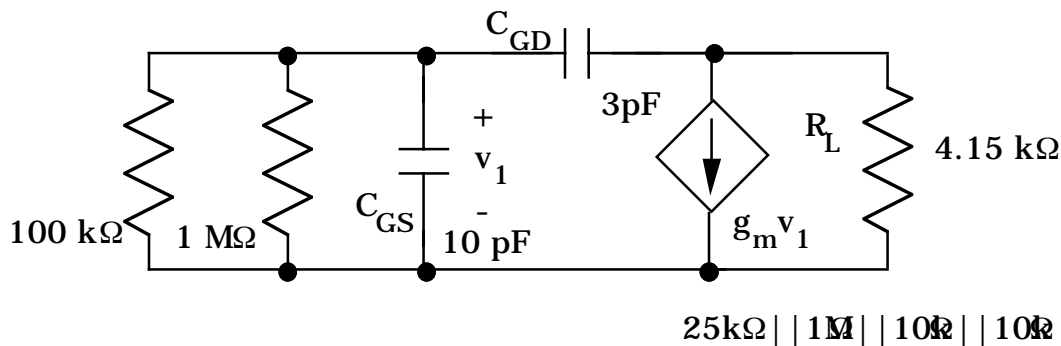
So the closed - loop values are $f_H = 101\text{ kHz}$ and $f_L = 0$.

5.) Problem 18.62 (18.30) of the text.



$$\omega_1 = \frac{1}{10^{-6}(100\text{k}\Omega + 1\text{M}\Omega)} = 0.909 \frac{\text{rad}}{\text{s}} \quad | \quad \omega_2 = \frac{1}{10^{-6}(10\text{k}\Omega + 25\text{k}\Omega \parallel 10\text{k}\Omega \parallel 1\text{M}\Omega)} = 58.5 \frac{\text{rad}}{\text{s}}$$

Separate, widely - spaced, poles $\rightarrow f_L^A = f_2 = \frac{58.5}{2\pi} = 9.31 \text{ Hz}$



$$\omega_H^A = \frac{1}{r_{\pi} C_T} = \frac{1}{(100\text{k}\Omega \parallel 1\text{M}\Omega) \left[10\text{pF} + 3\text{pF} \left(1 + 2\text{mS}(4.15\text{k}\Omega) + \frac{4.15\text{k}\Omega}{100\text{k}\Omega \parallel 1\text{M}\Omega} \right) \right]}$$

$$f_H^A = \frac{1}{2\pi} \frac{1}{(90.9\text{k}\Omega)(38.0\text{pF})} = 46.1 \text{ kHz}$$

$$v_{gs} = i_s (100\text{k}\Omega \parallel 1\text{M}\Omega) = (90.9\text{k}\Omega) i_s \quad | \quad v_o = -(2 \times 10^{-3}) v_{gs} (25\text{k}\Omega \parallel 10\text{k}\Omega \parallel 10\text{k}\Omega \parallel 1\text{M}\Omega)$$

$$A = \frac{v_o}{i_s} = -(2\text{mS})(4.15\text{k}\Omega)(90.9\text{k}\Omega) = -7.55 \times 10^5 \Omega \quad | \quad y_{12}^F = -10^{-5} \text{ S}$$

$$1 + A\beta = 1 + (-7.55 \times 10^5 \Omega)(-10^{-6} \text{ S}) = 1.76$$

$$f_L = \frac{9.31}{1.76} = 5.29 \text{ Hz} \quad f_H = 46.1\text{kHz}(1.76) = 81.0 \text{ kHz}$$