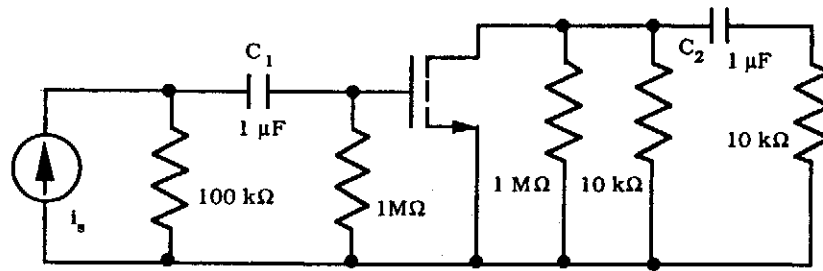
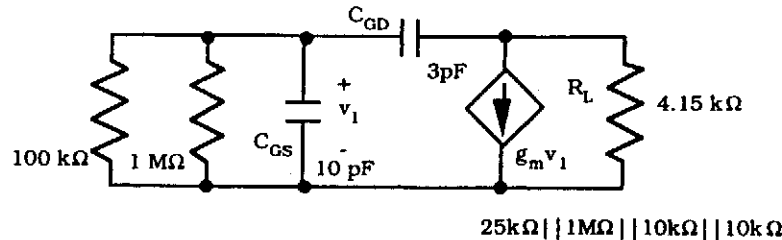


Homework Assignment No. 14 Solutions**18.30**

$$\omega_1 = \frac{1}{10^{-6}(100\text{k}\Omega + 1\text{M}\Omega)} = 0.909 \frac{\text{rad}}{\text{s}} \quad | \quad \omega_2 = \frac{1}{10^{-6}(10\text{k}\Omega + 25\text{k}\Omega \parallel 10\text{k}\Omega \parallel 1\text{M}\Omega)} = 58.5 \frac{\text{rad}}{\text{s}}$$

$$\text{Separate widely spaced poles} \rightarrow f_L^A = f_2 = \frac{58.5}{2\pi} = 9.31 \text{ Hz}$$



$$\omega_H^A = \frac{1}{r_{\pi o} C_T} = \frac{1}{(100\text{k}\Omega \parallel 1\text{M}\Omega) \left[10\text{pF} + 3\text{pF} \left(1 + 2\text{mS}(4.15\text{k}\Omega) + \frac{4.15\text{k}\Omega}{100\text{k}\Omega \parallel 1\text{M}\Omega} \right) \right]}$$

$$f_H^A = \frac{1}{2\pi (90.9\text{k}\Omega)(38.0\text{pF})} = 46.1 \text{ kHz}$$

$$\mathbf{v}_{gs} = \mathbf{i}_s (100\text{k}\Omega \parallel 1\text{M}\Omega) = (90.9\text{k}\Omega) \mathbf{i}_s \quad | \quad \mathbf{v}_o = -(2 \times 10^{-3}) \mathbf{v}_{gs} (25\text{k}\Omega \parallel 10\text{k}\Omega \parallel 10\text{k}\Omega \parallel 1\text{M}\Omega)$$

$$A = \frac{\mathbf{v}_o}{\mathbf{i}_s} = -(2\text{mS})(4.15\text{k}\Omega)(90.9\text{k}\Omega) = -7.55 \times 10^5 \Omega \quad | \quad y_{12}^F = -10^{-5} \text{ S}$$

$$1 + A\beta = 1 + (-7.55 \times 10^5 \Omega)(-10^{-6} \text{ S}) = 1.76$$

$$f_L = \frac{9.31}{1.76} = 5.29 \text{ Hz} \quad f_H = 46.1\text{kHz}(1.76) = 81.0 \text{ kHz}$$

18.32

$$(a) A(s) = \frac{2 \times 10^{14} \pi^2}{(2\pi \times 10^3)(2\pi \times 10^5)} = \frac{5 \times 10^5}{\left(1 + \frac{s}{2\pi \times 10^3}\right) \left(1 + \frac{s}{2\pi \times 10^5}\right)} = \frac{5 \times 10^5}{\left(1 + \frac{s}{2\pi \times 10^3}\right) \left(1 + \frac{s}{2\pi \times 10^5}\right)}$$

A(s) represents a low - pass amplifier with two widely - spaced poles

$$\text{Open - loop: } A_o = 5 \times 10^5 = 114\text{dB} \quad | \quad f_L = 0 \quad | \quad f_H = f_1 = 1000 \text{ Hz}$$

(b) A common mistake would be the following:

$$\text{Closed - loop: } f_H = 1000\text{Hz} \left[1 + 5 \times 10^5 (0.01) \right] = 5\text{MHz}$$

Oops! - This exceeds $f_2 = 100 \text{ kHz}$! This is a two - pole amplifier.

Prob. 18.32 - Cont'd

$$A_V(s) = \frac{\frac{2 \times 10^{14} \pi^2}{(s + 2\pi \times 10^3)(s + 2\pi \times 10^5)}}{1 + \frac{2 \times 10^{14} \pi^2}{(s + 2\pi \times 10^3)(s + 2\pi \times 10^5)}} (0.01) = \frac{2 \times 10^{14} \pi^2}{s^2 + 1.01(2\pi \times 10^5)s + 2 \times 10^{12} \pi^2}$$

Using dominant - root factorization: $f_1 = 101 \text{ kHz}$, $f_2 = 4.95 \text{ MHz}$

So the closed - loop values are $f_H = 101 \text{ kHz}$ and $f_L = 0$.

18.35

$$T = \frac{v_o}{v_x} = g_{m2}(r_{o2} \parallel r_{o4}) \frac{(\beta_o + 1)R}{(r_{o2} \parallel r_{o4}) + r_{\pi 3} + (\beta_o + 1)R} \quad | \quad g_{m1} = 40(10^{-4}) = 4.00 \text{ mS}$$

$$r_{o2} = \frac{50 + 1.4}{10^{-4}} = 514 \text{ k}\Omega \quad | \quad r_{o4} = \frac{50 + 11.3}{10^{-4}} = 613 \text{ k}\Omega \quad | \quad r_{\pi 3} = \frac{100(0.025)}{(12 \text{ V} / 10 \text{ k}\Omega)} = 2.08 \text{ k}\Omega$$

$$T = (4 \times 10^{-3})(280 \text{ k}\Omega) \frac{(101)10 \text{ k}\Omega}{280 \text{ k}\Omega + 2.08 \text{ k}\Omega + 101(10 \text{ k}\Omega)} = 876 \quad (58.9 \text{ dB})$$

18.49

(a) $T = A\beta = \frac{2 \times 10^{14} \pi^2}{(s + 2 \times 10^3 \pi)(s + 2 \times 10^5 \pi)} \left(\frac{1}{5} \right)$ | Yes, it is a second - order system and will have some phase margin, although Φ_M may be vanishingly small.

(b) For $\omega \gg 2\pi \times 10^5$, $|T(j\omega)| \approx \frac{4 \times 10^{13} \pi^2}{\omega^2}$ and $|T(j\omega)| = 1$ for $\omega = 1.987 \times 10^7 \frac{\text{rad}}{\text{s}}$

$\angle T(j1.987 \times 10^7) = -\tan^{-1} \frac{1.987 \times 10^7}{2000\pi} - \tan^{-1} \frac{1.987 \times 10^7}{2 \times 10^5 \pi} = 178.2^\circ \rightarrow \Phi_M = 1.83^\circ$ | A very small phase margin.

18.61

$$A_{V1} = \frac{V_{o1}}{V_{o2}} = -\frac{1}{sRC} \quad V_{o2} = \left(1 + \frac{2R}{2R} \right) V_+ = 2V_+$$

$$(V_+ - V_{o1}) \frac{G}{2} + sCV_+ + (V_+ - V_{o2})G_F = 0 \quad \text{Combining these yields}$$

$$A_{V2} = \frac{V_{o2}}{V_{o1}} = \frac{G}{sC + \left(\frac{G}{2} - G_F \right)} \quad \text{and} \quad T(s) = A_{V1}A_{V2} = \frac{1}{sRC \left(sRC + \frac{1}{2} - \frac{R}{R_F} \right)}$$

$$\angle T(j\omega_o) = 0 \rightarrow R_F = 2R \quad \text{and} \quad |T(j\omega_o)| = 1 \rightarrow \omega_o = \frac{1}{RC}$$
